# Mobile Communications TCS 455

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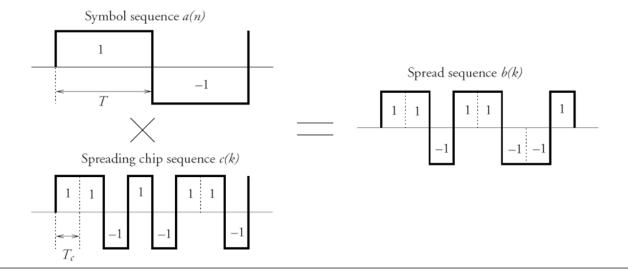
Office Hours: BKD 3601-7 Tuesday 14:00-16:00 Thursday 9:30-11:30

#### Announcements

- Read
  - Chapter 9: 9.1 9.5
- HW5 is posted.
  - Due: Feb 5 (This Friday)

### **DSSS and Maximum-length Codes**

- Maximal length codes have **excellent auto-correlation** properties (for ISI rejection), they have a number of properties that make them highly suboptimal for exploiting the multiuser capabilities of spread spectrum.
- There are only a small number of maximal length codes of a given length.
- Moreover, maximal length codes generally have relatively poor cross-correlation properties, at least for some sets of codes.





- For spread spectrum systems with **multiple users**, codes such as Gold, Kasami, or Walsh codes are used instead of maximal length codes
- Superior cross-correlation properties.
- Worse auto-correlation than maximal length codes.
  - The autocorrelation function of the spreading code determines its multipath rejection properties.

## Orthogonality

- Two vectors/functions are orthogonal if their inner product is zero.
- The symbol  $\perp$  is used to denote orthogonality.

Vector:  

$$\left\langle \vec{a}, \vec{b} \right\rangle = \vec{a} \cdot \vec{b}^{*} = \begin{pmatrix} a_{1} \\ \vdots \\ a_{n} \end{pmatrix} \cdot \begin{pmatrix} b_{1} \\ \vdots \\ b_{n} \end{pmatrix}^{*} = \sum_{k=1}^{n} a_{k} b_{k}^{*} = 0$$
Time-domain:  

$$\left\langle a, b \right\rangle = \int_{-\infty}^{\infty} a(t) b^{*}(t) dt = 0$$
Frequency domain:  

$$\left\langle A, B \right\rangle = \int_{-\infty}^{\infty} A(f) B^{*}(f) df = 0$$
Example (Fourier Series):  

$$\sin\left(2\pi k_{1} \frac{t}{T}\right) \text{ and } \cos\left(2\pi k_{2} \frac{t}{T}\right) \text{ on } [0,T]$$

$$e^{j2\pi n\frac{t}{T}} \text{ on } [0,T]$$

#### Parseval's Theorem

$$\langle x, y \rangle = \int_{-\infty}^{\infty} x(t) y^{*}(t) dt = \int_{-\infty}^{\infty} X(f) Y^{*}(f) df = \langle X, Y \rangle$$

If  $x(t) \perp y(t)$ , then  $X(f) \perp Y(f)$ .

#### **Orthogonality in Communication**

 $s(t) = \sum_{k=0}^{\ell-1} S_k c(t - kT_s) \xrightarrow{\mathcal{F}} S(f) = C(f) \sum_{k=0}^{\ell-1} S_k e^{-j2\pi f kT_s}$ 

 $s(t) = \sum_{k=0}^{\ell-1} S_k c_k(t) \xrightarrow{\mathcal{F}} S(f) = \sum_{k=0}^{\ell-1} S_k C_k(f) \quad \text{where} \quad c_{k_1} \perp c_{k_2}$ 

where 
$$c(t)$$
 is time-limited to  $[0, T]$ .

This is a special case of CDMA with  $c_k(t) = c(t - kT_s)$ 

The  $c_k$  are non-overlapping in time domain.



CDMA

TDMA

$$S(f) = \sum_{k=0}^{\ell-1} S_k C(f - k\Delta f)$$

where C(f) is frequency-limited to  $[0,\Delta f]$ . This is a special case of CDMA with  $C_k(f) = C(f - k\Delta f)$ 

The  $C_k$  are non-overlapping in freq. domain.

# Chapter 4 Multiple Access

Synchronous CDMA

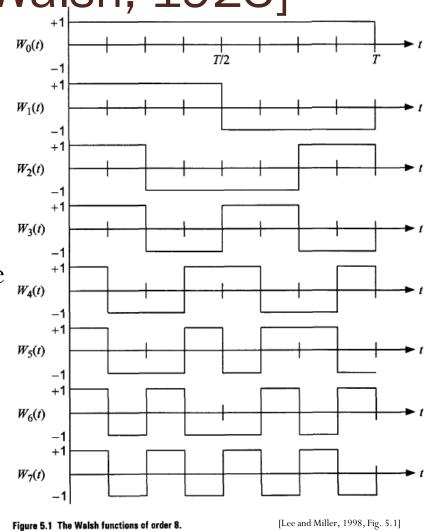
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### Synchronous CDMA Model

- Timing is important for orthogonality
- It is not possible to obtain orthogonal codes for asynchronous users.
- For synchronous users there is only a finite number of spreading codes that are orthogonal within any given bandwidth.
- Bit epochs are <u>aligned at the receiver</u>
- Require
  - Closed-loop timing control or
  - Providing the transmitters with access to a common clock (such as the Global Positioning System)

## Walsh Functions [Walsh, 1923]

- Walsh codes are used in second- (2G) and thirdgeneration (3G) cellular radio systems for providing channelization
- A set of Walsh functions can be ordered according to the number of zero crossing (sign changes)



#### Walsh Functions (2)

We define the Walsh functions of order N as a set of N time functions, denoted  $\{W_j(t); t \in (0,T), j = 0, 1, ..., N-1\}$ , such that

- W<sub>j</sub>(t) takes on the values {+1, −1} except at the jumps, where it takes the value zero.
- $W_j(0) = 1$  for all j.
- W<sub>j</sub>(t) has precisely j sign changes (zero crossings) in the interval (0, T).
- $\int_0^T W_j(t) W_k(t) dt = \begin{cases} 0, & \text{if } j \neq k \\ T, & \text{if } j = k \end{cases}$  Orthogonality
- Each function W<sub>j</sub>(t) is either odd or even with respect to the midpoint of the interval.

Once we know how to generate these Walsh functions of any order *N*, we can use them in *N*-channel orthogonal multiplexing applications.

#### Walsh Sequences

#### Walsh sequences

	_					_	-						_		
$W_0 = 0$	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
$W_1 = 0$	0	0	0	0	0	0	0	1	1	1	1	1	1	1	1
$oldsymbol{W}_2=0$	0	0	0	1	1	1	1	1	1	1	1	0	0	0	0
$W_{3} = 0$	0	0	0	1	1	1	1	0	0	0	0	1	1	1	1
$W_4 = 0$	0	1	1	1	1	0	0	0	0	1	1	1	1	0	0
$W_{5} = 0$	0	1	1	1	1	0	0	1	1	0	0	0	0	1	1
$W_{6} = 0$	0	1	1	0	0	1	1	1	1	0	0	1	1	0	0
$W_{7} = 0$	0	1	1	0	0	1	1	0	0	1	1	0	0	1	1
$oldsymbol{W}_8=0$	1	1	0	0	1	1	0	0	1	1	0	0	1	1	0
$W_{9} = 0$	1	1	0	0	1	1	0	1	0	0	1	1	0	0	1
$W_{10} = 0$	1	1	0	1	0	0	1	1	0	0	1	0	1	1	0
$W_{11} = 0$	1	1	0	1	0	0	1	0	1	1	0	1	0	0	1
$W_{12} = 0$	1	0	1	1	0	1	0	0	1	0	1	1	0	1	0
$W_{13} = 0$	1	0	1	1	0	1	0	1	0	1	0	0	1	0	1
$W_{14} = 0$	1	0	1	0	1	0	1	1	0	1	0	1	0	1	0
$W_{15} = 0$	1	0	1	0	1	0	1	0	1	0	1	0	1	0	1

- The Walsh functions, expressed in terms of ±1 values, form a group under the multiplication operation (multiplicative group).
- The Walsh sequences, expressed in terms of (0, 1) values, form a group under modulo-2 addition (additive group).
- Closure property:  $W_i(t) \cdot W_j(t) = W_r(t)$  $W_i \oplus W_i = W_r$

#### Walsh sequences of order 64

#### Table 5.2 Walsh functions of order 64 (indexed by zero crossings)

$W_0$	00000000000000 000000000000 00000000000
W <sub>1</sub>	00000000000000 000000000000000000000000
$W_2$	000000000000000000000000000000000000000
$W_3$	000000000000000000000000000000000000000
$W_4$	000000011111111 111111100000000 0000000111111
W.	0000000011111111 111111100000000 1111111
$W_6$	0000000011111111 0000000011111111 111111
$W_7$	0000000011111111 00000000111111111 000000
$W_8$	0000111111110000 0000111111110000 0000111111
W	0000111111110000 0000111111110000 1111000000
$W_{10}$	0000111111110000 1111000000001111 1111000000
Wi	0000111111110000 1111000000001111 0000111111
$W_{12}^{-1}$	0000111100001111 1111000011110000 0000111100001111 1111000011110000
$W_{13}$	0000111100001111 1111000011110000 1111000011110000 0000111100001111
$W_{14}$	0000111100001111 0000111100001111 1111000011110000 1111000011110000
$W_{15}$	$0000111100001111\ 0000111100001111\ 0000111100001111\ 0000111100001111$
$W_{16}$	0011110000111100 0011110000111100 0011110000111100 0011110000111100
$W_{17}$	0011110000111100 0011110000111100 1100001111000011 1100001111000011
$W_{18}$	0011110000111100 1100001111000011 1100001111000011 0011110000111100
$W_{19}$	0011110000111100 1100001111000011 0011110000111100 1100001111000011
$W_{20}$	0011110011000011 1100001100111100 0011110011000011 1100001100111100
$W_{21}$	0011110011000011 1100001100111100 1100001100111100 0011110011000011
$W_{22}$	0011110011000011 0011110011000011 1100001100111100 1100001100111100
$W_{23}$	0011110011000011 0011110011000011 0011110011000011 0011110011000011
$W_{24}$	0011001111001100 0011001111001100 0011001111001100 0011001111001100
$W_{25}$	0011001111001100 0011001111001100 1100110000110011 1100110000110011
$W_{26}$	0011001111001100 1100110000110011 1100110000110011 0011001111001100
$W_{27}$	0011001111001100 1100110000110011 0011001111001100 1100110000110011
$W_{28}$	0011001100110011 1100110011001100 0011001100110011 1100110011001100
$W_{29}$	0011001100110011 1100110011001100 1100110011001100 0011001100110011
$W_{30}$	0011001100110011 001100110011 110011001
$W_{31}$	0011001100110011 0011001100110011 0011001100110011 0011001100110011

W32 0110011001100110 0110011001100110 0110011001100110 0110011001100110 W33 0110011001100110 0110011001100110 1001100110011001 1001100110011001  $W_{34}$ 0110011001100110 1001100110011001 1001100110011001 0110011001100110  $W_{35}$ 0110011001100110 1001100110011001 0110011001100110 100110011001  $W_{36}$ 0110011010011001 1001100101100110 0110011010011001 1001100101100110  $W_{37}$ 0110011010011001 1001100101100110 1001100101100110 0110011010011001  $W_{38}$ 0110011010011001 0110011010011001 1001100101100110 1001100101100110 W39 0110011010011001 0110011010011001 0110011010011001 0110011010011001  $W_{40}$ 0110100110010110 0110100110010110 0110100110010110 0110100110010110  $W_{41}$ 0110100110010110 0110100110010110 100101100110011 1001011001101001 W43 0110100110010110 1001011001101001 0110100110010110 1001011001101001  $W_{44}$  $W_{45}$  $W_{46}$  $W_{47}$  $W_{48}$  $W_{49}$  $W_{50}$  $W_{51}$  $W_{52}$  $W_{53}$  $W_{54}$  $W_{55}$  $W_{56}$ 010101010101010 010101010101010 01010101010101010 01010101010101010  $W_{57}$  $W_{58}$ 010101010101010 1010101001010101 1010100001010101 010101010101010  $W_{50}$ 010101010101010 1010101001010101 01010101010101010 10101010101010101  $W_{60}$  $W_{61}$  $W_{62}$  $W_{63}$ 

#### Walsh Function Generation

- The Walsh functions can be generated (or computed) by many methods. We can construct the Walsh functions by:
  - Using Rademacher functions;
  - Using Hadamard matrices;
  - Exploiting the symmetry properties of Walsh functions themselves.
- The **Hadamard matrix** is a square array of plus and minus ones, {+1, -1}, whose rows and columns are mutually orthogonal.
- If the first row and first column contain only plus ones, the matrix is said to be in **normal form**.
- We can replace "+1" with "0" and "-1" with "1" to express the Hadamard matrix using the logic elements (0, 1).
- The 2×2 Hadamard matrix of order 2 is

$$H_2 = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} \equiv \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix}$$

#### Hadamard matrix (1)

Suppose  $H_N$  is an  $N \times N$  Hadamard matrix.  $N \ge 1$  is called the order of a Hadamard matrix

1. N = 1, 2, or 4t, where t is a positive integer.

2.  $H_N H_N^T = N I_N$  where  $I_N$  is the  $N \times N$  identity matrix.

3. If  $H_a$  and  $H_b$  are Hadamard matrices of order a and b, respectively, then we define  $H_a \otimes H_b$  to be the Hadamard matrix  $H_{ab}$  of order ab whose elements are found by substituting  $H_b$  for +1 (or logic 0) in  $H_a$  and  $-H_b$  (or the complement of  $H_b$ ) for -1 (or logic 1) in  $H_a$ .

> Caution: Some textbooks write this symbol as ×. It is not the regular matrix multiplication

#### Hadamard matrix (2)

• Consequently, if N is a power of two and it is understood that  $H_1 = [+1] \equiv [0]$ , then  $H_{2N}$  can be found as follows:

$$H_{2N} = \left[ \begin{array}{cc} H_N & H_N \\ H_N & \overline{H_N} \end{array} \right]$$

where  $\overline{H_N}$  is the negative (complement) of  $H_N$ .

► Hadamard matrices of order N = 2<sup>t</sup> can be formed by repeatedly multiplying (⊗) the normal form of the N = 2 Hadamard matrix by itself.

#### Hadamard matrix: Examples

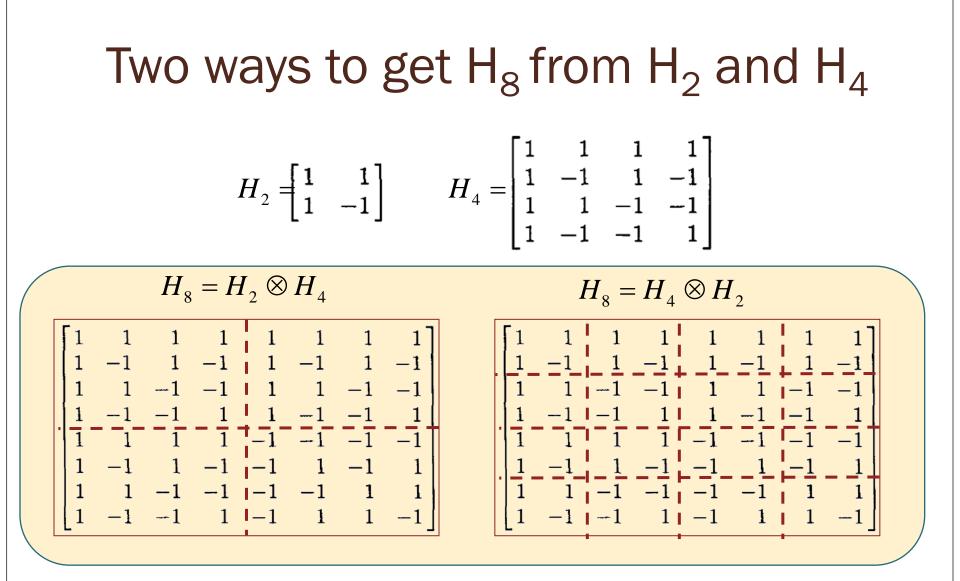
$$\mathbf{H_2} = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$
Caution: This is not the usual matrix multiplication!

$$\mathbf{H}_{16} = \mathbf{H}_2 \times \mathbf{H}_8 = \mathbf{H}_2$$

Γ1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	٦٦
i	$-1^{1}$	1	$-1^{1}$	-	-1	1		1		1		_	-	1	1
1	1	-1	-1	1	1		-1	1	-1			1	1	_	-1
	_		-	_	_	_	_	_				-			-1
1	-1	_	1	1	-1	$^{-1}$	1	1			1		$^{-1}$	_	L I
1	1	1	1	-1	-1	-1	-1	1	1	1	1	$^{-1}$	-1	-1	-1
1	-1	1	-1	-1	1	-1	1	1	-1	1	$^{-1}$	-1	1	-1	1
1	1	-1	$^{-1}$	-1	-1	1	1	1	1	-1	$^{-1}$	-1	-1	1	1
1	-1	-1	1	-1	1	1	-1	1	-1	-1	1	-1	1	1	-1
1	1	1	1	1	1	1	1	-1	-1	-1	-1	-1	-1	-1	-1
I .					_	_	_	_	-	-	-	-			_
[1	-1	1	-1	1	-1	1	-1	-1	1						1
$\begin{vmatrix} 1 \\ 1 \end{vmatrix}$	_	1 -1		1 1	-1 1		$-1 \\ -1$		1			-1	1 -1		1 1
$\begin{vmatrix} 1 \\ 1 \\ 1 \end{vmatrix}$	_	-1		-	_	-1		-1	1	-1	1	-1 -1	1	-1	1 1 -1
1 1 1 1	1	-1	-1	1	1	$-1 \\ -1$	$^{-1}$	- <b>1</b> -1	1 -1 1	$-1 \\ 1$	1 1 -1	-1 -1	$\frac{1}{-1}$	-1 1	1
1 1 1 1	1	$-1 \\ -1$	-1 1	1 1	1 1	$-1 \\ -1$	$^{-1}_{1}$	-1 -1 -1	1 -1 1	-1 1 1	1 1 -1	$-1 \\ -1 \\ -1$	$\frac{1}{-1}$	-1 1	1
1 1 1 1 1 1	1 -1 1 -1	$-1 \\ -1 \\ 1$	-1 1 1 -1	1 1 -1	1 1 1	-1 -1 -1	-1 1 -1 1	-1 -1 -1	$     \begin{array}{c}       1 \\       -1 \\       1 \\       -1 \\       1     \end{array} $	-1 1 1 -1	1 1 -1 -1	$-1 \\ -1 \\ -1 \\ 1$	1 1 1 I	-1 1 1 1	1 1 -1 1

In MATLAB, use hadamard(k)

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#### Walsh-Hadamard Sequences

- All the row (or column) sequences of Hadamard matrices are Walsh sequences if the order is  $N = 2^{t}$ .
- The Walsh functions generated by the Hadamard matrix method are not indexed according to the number of sign changes.
- Used in synchronous CDMA
  - It is possible to synchronize users on the downlink, where all signals originate from the same transmitter.
  - It is more challenging to synchronize users in the uplink, since they are not co-located.
    - Asynchronous CDMA

## Hadamard Matrix in MATLAB

• We use the hadamard function in MATLAB to generate Walsh functions of length eight.

N = 8; % Length of Walsh (Hadamard) functions
hadamardMatrix = hadamard(N)
hadamardMatrix =

1	1	1	1	1	1	1	1
1	-1	1	-1	1	-1	1	-1
1	1	-1	-1	1	1	-1	-1
1	-1	-1	1	1	-1	-1	1
1	1	1	1	-1	-1	-1	-1
1	-1	1	-1	-1	1	-1	1
1	1	-1	-1	-1	-1	1	1
1	-1	-1	1	-1	1	1	-1

• The Walsh functions in the matrix are not arranged in increasing order of their sequencies or number of zero-crossings (i.e. 'sequency order') .

#### Walsh Matrix in MATLAB

• The Walsh matrix, which contains the Walsh functions along the rows or columns in the increasing order of their sequencies is obtained by changing the index of the hadamardMatrix as follows.

```
HadIdx = 0:N-1;
M = log2(N)+1;
```

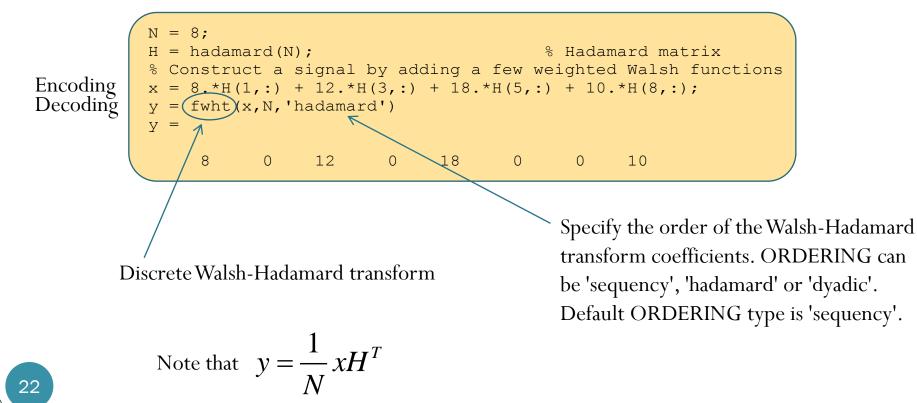
% Hadamard index % Number of bits to represent the index

• Each column of the sequency index (in binary format) is given by the modulo-2 addition of columns of the bit-reversed Hadamard index (in binary format).

```
binHadIdx = fliplr(dec2bin(HadIdx,M)); % Bit reversing of the binary index
binHadIdx = uint8(binHadIdx)-uint8('0'); % Convert from char to integer array
binSeqIdx = zeros(N,M-1,'uint8'); % Pre-allocate memory
for k = M:-1:2
   % Binary sequency index
   binSeqIdx(:,k) = xor(binHadIdx(:,k),binHadIdx(:,k-1));
end
SeqIdx = bin2dec(int2str(binSeqIdx)); % Binary to integer sequency index
walshMatrix = hadamardMatrix(SeqIdx+1,:) % 1-based indexing
walshMatrix =
    1
                                         1
        1
    1 1
1 1
    1
    1
    1
    1
        -1
           1
                 -1 1
                              -1
                                1
                                        -1
```

#### **CDMA via Hadamard Matrix**

• This signal is formed using weighted Walsh functions, so the WHT should return non-zero values equal to the weights at the respective indices



# Chapter 4 Multiple Access

IS-95

Office Hours: BKD 3601-7 Tuesday 14:00-16:00 Thursday 9:30-11:30

#### Evolution of cellular network

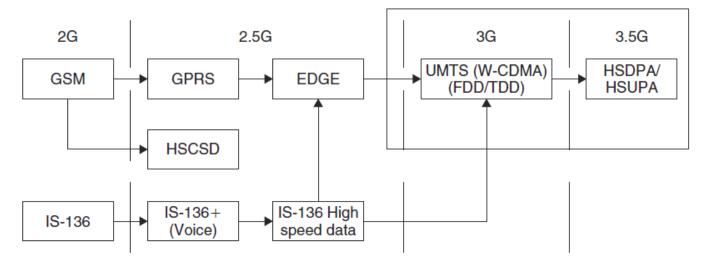


Figure 1.1 Evolution of 2G networks based on TDMA technology.

[Abu-Rgheff, 2007]

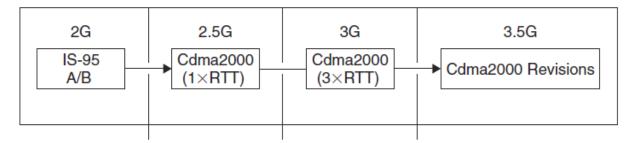


Figure 1.2 Evolution of 2G networks based on CDMA technology.

## IS-95 System commone

- Based on direct sequence CDMA (DS-CDMA)
  - First CDMA-based digital cellular standard.
- The brand name for IS-95 is cdmaOne.
  - Also known as TIA-EIA-95.
- Proposed by Qualcomm in 1989 and adopted in 1933.
  - North America
- Now being replaced by IS-2000 (CDMA2000)
- 1.25 MHz Channel BW
- 1.228 Mb/s chip rate
- Walsh functions of "order 64" are extensively used in the IS-95 system.
- Remarks
  - IS-95B = cdmaOne
    - Ugrade IS-95A
    - Can carry data at rates up to 14.4 kbps for IS-95A and 115 kbps for IS-95B.



#### 64-ary Walsh Functions

 $egin{array}{c} {m W}_1 \ {m W}_{62} \ {m W}_{30} \ {m W}_{33} \end{array}$ 

 $W_{14}$  $W_{49}$  $W_{17}$  $W_{46}$ 

 $W_{34}$  $W_{13}$  $W_{50}$  $W_{18}$ 

 $W_{45}$  $W_5$ 

 $egin{array}{c} W_{58} \ W_{26} \ W_{37} \ W_{10} \ W_{53} \ W_{21} \ W_{12} \end{array}$ 

#### Table 5.8 Walsh functions of order 64, as indexed in IS-95 ( $W_i$ is the Walsh notation, and $H_i$ is the Hadamard notation)

$W_0$	$H_0$	000000000000000000000000000000000000000
$W_{63}$	$H_1$	0101010101010101 0101010101010101 010101010101010101 010101010101010101
$W_{31}$	$H_2$	0011001100110011 0011001100110011 0011001100110011 001100110011
$W_{32}$	$H_3$	0110011001100110 011001100110 011001100
$W_{15}$	$H_4$	0000111100001111 0000111100001111 0000111100001111 0000111100001111
$W_{48}$	$H_5$	0101101001011010 0101101001011010 0101101
$W_{16}$	$H_6$	0011110000111100 0011110000111100 0011110000111100 0011110000111100
$W_{47}$	$H_7$	0110100101101001 0110100101101001 011010010
$W_7$	$H_8$	0000000011111111 0000000011111111 000000
$W_{56}$	$H_9$	010101010101010 0101010110101010 01010101010101010 01010101010101010
$W_{24}$	$H_{10}$	0011001111001100 0011001111001100 0011001111001100 0011001111001100
$W_{39}$	$H_{11}$	0110011010011001 0110011010011001 0110011010011001 0110011010011001
$W_8$	$H_{12}$	0000111111110000 0000111111110000 0000111111
$W_{55}$	$H_{13}$	0101101010100101 0101101010100101 0101101
$W_{23}$	$H_{14}$	0011110011000011 0011110011000011 0011110011000011 0011110011000011
$W_{40}$	$H_{15}$	0110100110010110 0110100110010110 0110100110010110 0110100110010110
$W_3$	$H_{16}$	000000000000000 11111111111111 00000000
$W_{60}$	$H_{17}$	0101010101010101 1010101010101010 010101010101010101 10
$W_{28}$	$H_{18}$	0011001100110011 1100110011001100 0011001100110011 1100110011001100
$W_{35}$	$H_{19}$	0110011001100110 1001100110011001 0110011001100110 1001100110011001
$W_{12}$	$H_{20}$	0000111100001111 1111000011110000 0000111100001111 1111000011110000
$W_{51}$	$H_{21}$	0101101001011010 1010010110100101 0101101
$W_{19}$	$H_{22}$	0011110000111100 1100001111000011 0011110000111100 1100001111000011
$W_{44}$	$H_{23}$	0110100101101001 1001011010010110 011010010
$W_4$	$H_{24}$	0000000011111111 111111100000000 0000000
$W_{59}$	$H_{25}$	0101010110101010 1010101001010101 01010101010101010 10
$W_{27}$	$H_{26}$	0011001111001100 1100110000110011 0011001111001100 1100110000110011
$W_{36}$	$H_{27}$	0110011010011001 1001100101100110 0110011010011001 1001100101100110
$W_{11}$	$H_{28}$	0000111111110000 1111000000001111 0000111111
$W_{52}$	$H_{29}$	0101101010100101 101001010101010 0101101
$W_{20}$	$H_{30}$	0011110011000011 1100001100111100 0011110011000011 1100001100111100
$W_{43}$	$H_{31}$	0110100110010110 1001011001101001 0110100110010110 1001011001101001

$H_{32}$	000000000000000000000000000000000000000
$H_{33}^{**}$	0101010101010101 0101010101010101 101010101010101010 1010101010101010
$H_{34}$	0011001100110011 0011001100110011 1100110011001100 1100110011001100
$H_{35}$	0110011001100110 0110011001100110 1001100110011001 1001100110011001
$H_{36}$	0000111100001111 0000111100001111 1111000011110000 1111000011110000
$H_{37}$	0101101001011010 0101101001011010 1010010
$H_{38}$	0011110000111100 0011110000111100 1100001111000011 1100001111000011
$H_{39}$	0110100101101001 0110100101101001 100101101
$H_{40}$	0000000011111111 0000000011111111 111111
$H_{41}$	010101010101010 010101010101010 10101010101010101 10101010010
$H_{42}$	0011001111001100 0011001111001100 1100110000110011 1100110000110011
$H_{43}$	0110011010011001 0110011010011001 1001100101100110 1001100101100110
$H_{44}$	0000111111110000 0000111111110000 1111000000
$H_{45}$	0101101010100101 0101101010100101 1010010
$H_{46}$	0011110011000011 0011110011000011 1100001100111100 1100001100111100
$oldsymbol{H}_{47}$	0110100110010110 0110100110010110 1001011001101001 1001011001101001
$H_{48}$	000000000000000000000000000000000000000
$H_{49}$	0101010101010101 1010101010101010 1010101010101010 01
$H_{50}$	0011001100110011 1100110011001100 1100110011001100 001100110011
$H_{51}$	0110011001100110 1001100110011001 1001100110011001 0110011001100110
$H_{52}$	0000111100001111 1111000011110000 1111000011110000 0000111100001111
$H_{53}$	0101101001011010 1010010110100101 1010010
$H_{54}$	0011110000111100 1100001111000011 1100001111000011 0011110000111100
$H_{55}$	0110100101101001 1001011010010110 100101101
$H_{56}$	0000000011111111 1111111100000000 111111
$H_{57}$	010101010101010 1010101001010101 10101010010
$H_{58}$	0011001111001100 1100110000110011 1100110000110011 0011001111001100
$H_{59}$	0110011010011001 1001100101100110 1001100101100110 0110011010011001
$H_{60}$	0000111111110000 1111000000001111 1111000000
$H_{61}$	0101101010100101 101001010101010 1010010
$H_{62}$	0011110011000011 1100001100111100 1100001100111100 0011110011000011
$H_{63}$	0110100110010110 1001011001101001 1001011001101001 0110100110010110

[Lee and Miller, 1998, Table 5.8]

### Walsh Sequences in IS-95

#### • Forward link

- QPSK with a chip rate of 1,228,800 per second.
- The **multiple access scheme** is accomplished by the use of 64-bit spreading orthogonal **Walsh sequences** (functions).
  - The (coded and interleaved) traffic channel signal symbols are multiplied with distinct repeating Walsh sequences that are assigned to each channel for the duration of the call.
- Every base stations is synchronized with a GPS receiver so transmissions are tightly controlled in time.

#### • Reverse link

 The Walsh sequences are employed as an orthogonal modulation code, which <u>depends only on the data pattern</u> (not channel), forming a 64-ary orthogonal modulation system.

#### IS-95

- The reverse link is subject to near-far effects.
- More powerful error correction is employed on the reverse link.
  - A rate 1/2 constraint length 9 convolutional code followed by an interleaver on the forward channel while
  - A rate 1/3 constraint length 9 convolutional code followed by an interleaver is used on the reverse link.
  - Interleaving is utilized to avoid large burst errors, which can be very detrimental to convolutional codes.
- Power control.
  - Use a subchannel on the forward link
  - Every 1.25 ms the base station receiver estimates the signal strength of the mobile unit.
  - If it is too high, the base transmits a 1 on the subchannel. If it is too low, it transmits a 0.
  - In this way, the mobile station adjusts its power every 1.25 ms as necessary so as to reduce interference to other users.

#### **IS-95: Increased Spectral Efficiency**

- Improve frequency reuse.
  - Narrow-band systems cannot use the same transmission frequency in adjacent cells because of the potential for interference.
  - CDMA has inherent resistance to interference.
    - N = 1 (theoretically)
    - Although users from adjacent cells will contribute to interference level, their contribution will be significantly less than the interference from the same cell users.
    - Frequency reuse efficiency increases by a factor of 4 to 6.
- When used to transmit voice signals, CDMA systems may exploit the fact that voice activity typically lies at somewhat less than 40%, thus reducing the amount of interference to 40% of its original value.

## QCELP

- Qualcomm code-excited linear prediction algorithm
- Used for voice encoding.
- The voice coder exploits gaps and pauses in speech.
- The data rate is variable.
- To keep the symbol rate constant, whenever the bit rate falls below the peak bit rate of 9600 kbit/s, repetition is used to fill the gaps.
  - For example, if the output of the voice coder (and subsequently the convolutional coder) falls to 2400 bit/s, the output is repeated three times before it is sent to the interleaver.
  - Takes advantage of this repetition time by reducing the output power during three out of the four identical symbols by at least 20 dB.
  - In this way, the multiple-access interference is reduced.
- This voice activity gating reduces interference and increases overall capacity.